Lecture 3: Power Dissipation

CSCE 6933/5933 Advanced Topics in VLSI Systems

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Outline of the Talk

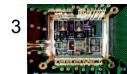
- Power and Energy
- Dynamic Power
- Static Power
- Low Power Design



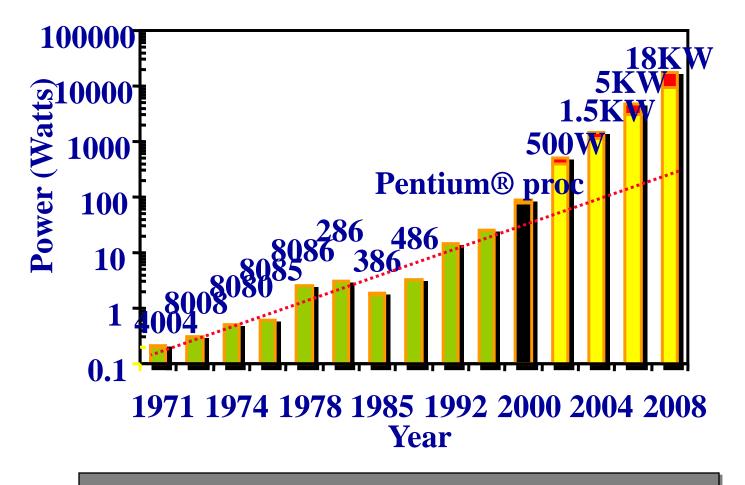


Power Dissipation Trend





Power Dissipation Trend

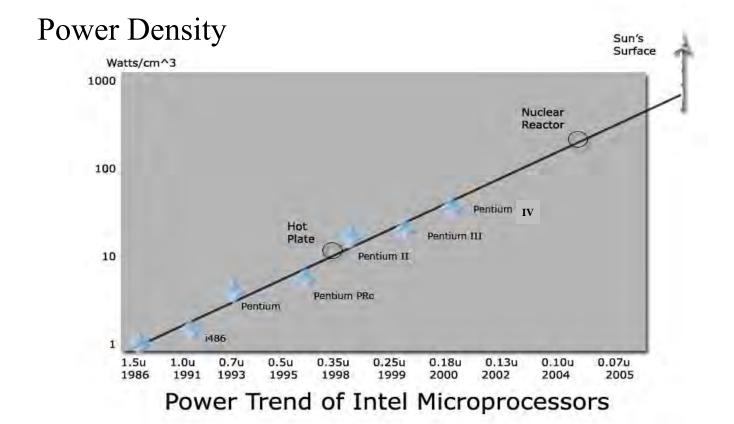


Power delivery and dissipation will be prohibitive



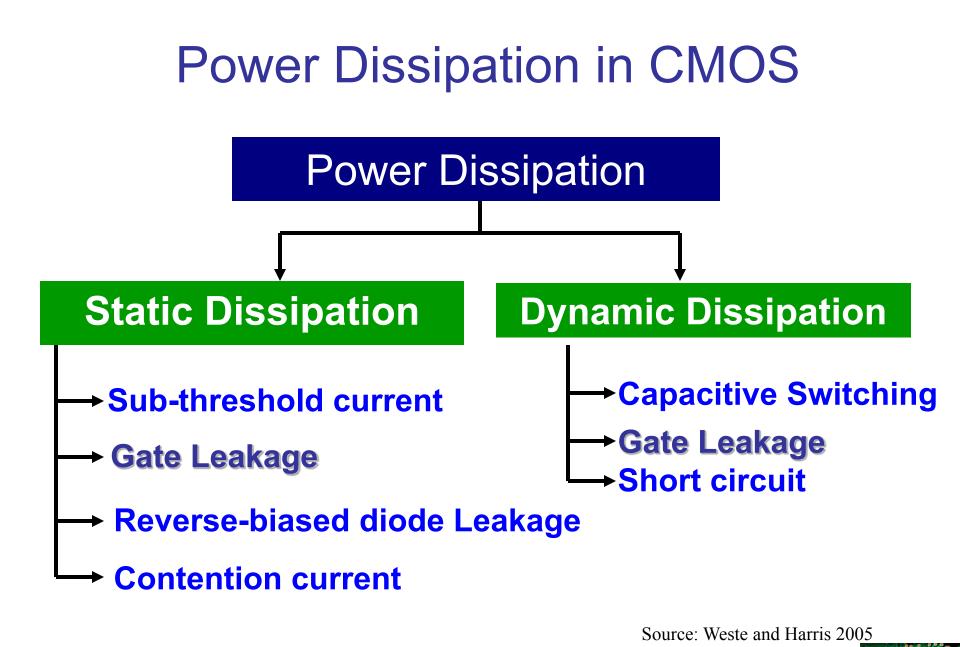


Why Low-Power?







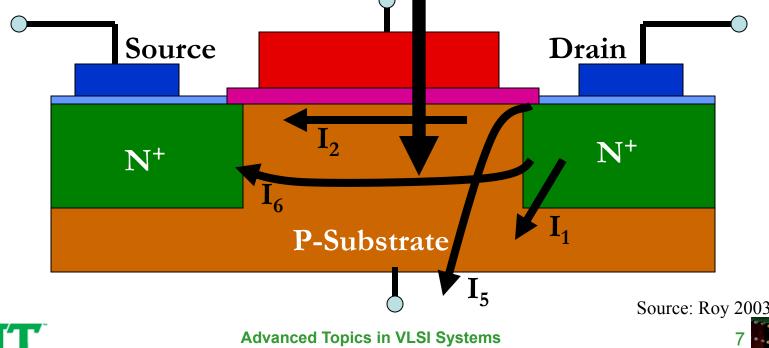




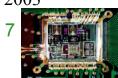
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Leakages in CMOS

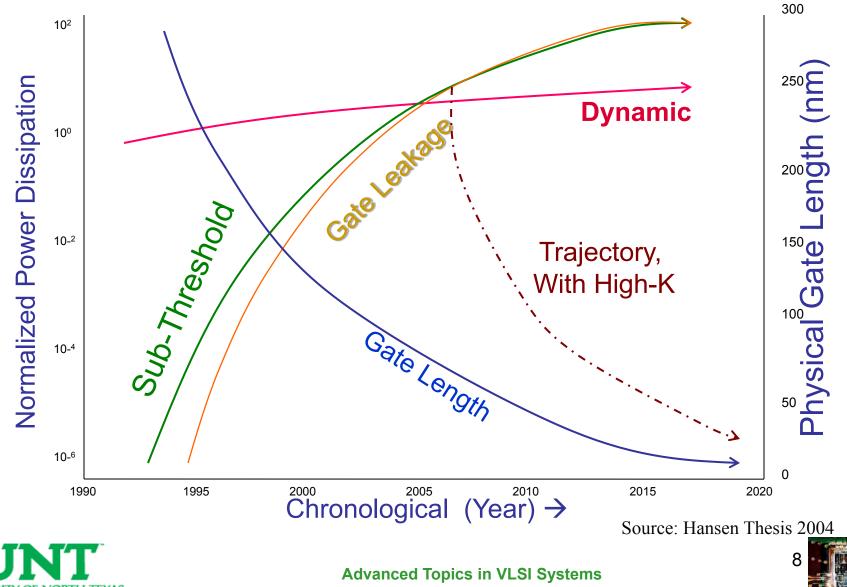
- I₁ : reverse bias pn junction (both ON & OFF)
- I₂: subthreshold leakage (OFF)
- I_3 : Gate Leakage current (both ON & OFF)
- I_4 : gate current due to hot carrier injection (both ON & OFF)
- I_5 : gate induced drain leakage (OFF)
- I₆ : channel punch through current (OFF)







Power Dissipation Redistribution



UNIVERSITY OF NORTH TEXAS Discover the power of ideas

Dynamic and Static Power Sources





Power Dissipation in CMOS : Dynamic

Capacitance Switching Current: This flows to charge and discharge capacitance loads during logic changes.

Short-Circuit Current: This is the current due to the DC path between the supply and ground during output transition.





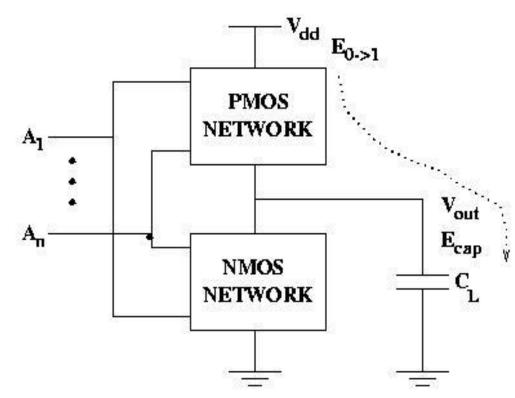
Power Dissipation in CMOS : Static

- Subthrehold Current: Sub-threshold current that arises from the inversion charges that exists at the gate voltages below the threshold voltage.
- Tunneling Current: There is a finite probability for carrier being pass through the gate oxide. This results in tunneling current thorough the gate oxide.
- Reverse-biased Diode Leakage: Reverse bias current in the parasitic diodes.
- Contention Current in Ratioed Circuits: Ratioed circuits burn power in fight between ON transistors





Power Dissipation in CMOS : Dynamic



A general CMOS transistor circuit

Discover the power of ideas

•Dynamic power is required to charge and discharge load capacitances when transistors switch.

- •One cycle involves a rising and falling output.
- •On rising output, charge Q = CLVDD is required.
- •On falling output, charge is dumped to GND.





Power Dissipation in CMOS : Dynamic

$$\mathbf{E}_{0\rightarrow 1} = \int_{0}^{T} \mathbf{P}(t) dt = \mathbf{V}_{dd} \int_{0}^{T} \mathbf{i}_{supply}(t) dt = \mathbf{V}_{dd} \int_{0}^{V_{dd}} \mathbf{C}_{L} d\mathbf{V}_{out} = \mathbf{C}_{L} \mathbf{V}_{dd}^{2}$$

$$E_{out} = \int_{0}^{T} P_{out}(t) dt = \int_{0}^{T} V_{out} i_{out}(t) dt = \int_{0}^{V_{dd}} C_{L} V_{out} dV_{out} = \frac{1}{2} C_{L} V_{dd}^{2}$$

Note:

- 1. the difference between the two is the loss
- 2. Energy doesn't depend on frequency





Power Dissipation in CMOS : Dynamic For N_c clock cycles energy loss :

$$\mathbf{E}_{\mathbf{N}_{\mathbf{C}}} = \mathbf{C}_{\mathbf{L}} \mathbf{V}_{\mathbf{d}\mathbf{d}}^2 \mathbf{n}(\mathbf{N}_{\mathbf{c}})$$

 $n(N_c)$: is the number of 0->1 transitions in N_c clock cycles

$$P_{avg} = \lim_{N \to inf} \left[\frac{E_{N_c}}{N_c} \right] f = \left[\lim_{N \to inf} \frac{n(N_c)}{N_c} \right] C_L V_{dd}^2 f$$
$$= \alpha_{0 > 1} C_L V_{dd}^2 f$$

Note: Power depends on frequency





Short Circuit Current

- When transistors switch, both NMOS and PMOS networks may be momentarily ON at once.
- Leads to a blip of "short circuit" current.
- < 10% of dynamic power if rise/fall times are comparable for input and output.





Static Power : Subthrehold Current

- In OFF state, undesired leakage current flow.
- It contributes to power dissipation of idle circuits.
- If v_t is the thermal voltage and I_0 is the current at V_{th} then the subthreshold current is :

$$I_{ds} = I_0 \left[1 - \exp\left(-\frac{V_{ds}}{v_t}\right) \right] \cdot \exp\left(\frac{V_{gs} - V_{th} - V_{off}'}{nv_t}\right) \qquad I_o = \mu C_{ox} \left(\frac{W}{L}\right) v_t^{2} e^{1.8}$$

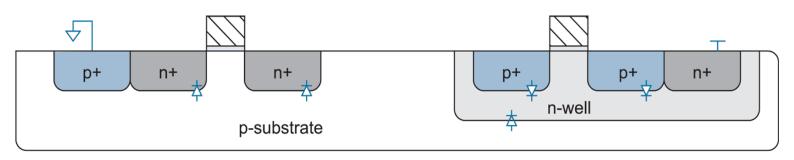
- Drain-Induced-Barrier-Lowering (DIBL) an prominent effect for short channel transistors also impacts subthreshold conduction by lowering V_{th} .
- It increases as the V_{th} decreases or V_{gs} increases.
- It increases as the temperature increases.





Static Power : Junction Leakage

- The pn junctions between diffusion, substrate and well are all junction diodes.
- These are revered biased as substrate is connected to GND and well connected to V_{dd} .
- However, reversed biased diode also conduct small amount of current.



Reverse-biased diodes in CMOS circuits





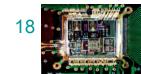
Static Power : Junction Leakage

 The reverse-biased junction current is expressed as follows: (D is not for drain, S is not for source)

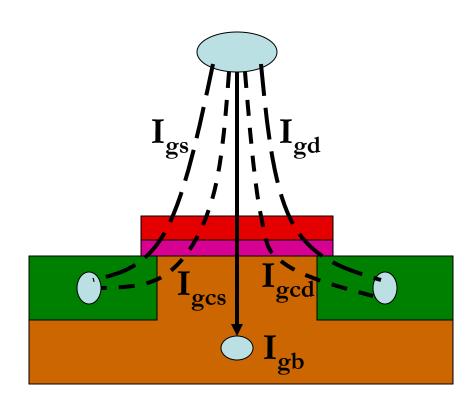
$$I_{\rm D} = I_{\rm S} \left[\exp \left(V_{\rm D} / v_{\rm T} \right) - 1 \right]$$

- I_S depends on the doping level, the area, and perimeter of the diffusion region.
- V_D is the diode voltage e.g. V_{sb} or V_{db} .





Static Power : Tunneling



BSIM4 Model

- •There is a finite probability for carrier being pass through the gate oxide.
- •This results in tunneling current thorough the gate oxide.
- •The effect is predominate for lower oxide thickness.





Static Power : Tunneling

 The gate oxide leakage current can be expressed as follows [Kim2003, Chandrakasan2001] (K and α are experimentally derived factors).

 $I_{gate} = K W_{gate} (V_{dd} / T_{gate})^2 \exp(-\alpha T_{gate} / V_{dd})$

- Options for reduction of gate leakage power :
 - Decreasing of supply voltage V_{dd} (will play its role)
 - Increasing gate SiO₂ thickness T_{gate} (opposed to the technology trend !!)
 - Decreasing gate width W_{gate} (only linearly dependent)



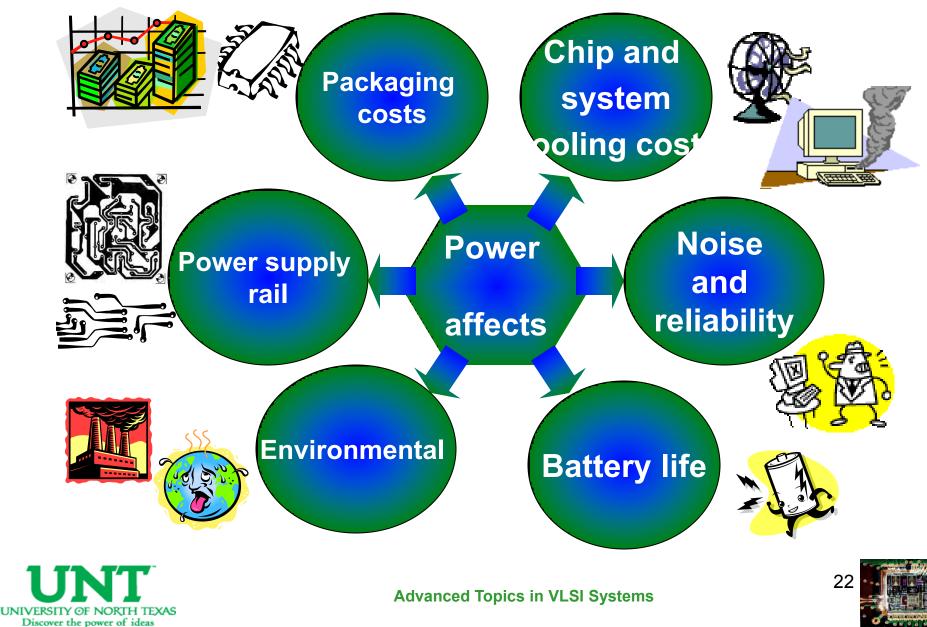


Low-Power Design





Why Low Power?



Various forms of Power Profile

- Average Power
- Total Energy
- Energy-Delay-Product (EDP)
- Power-Delay-Product (PDP)
- Power-Square-Delay-Product (PSDP)
- Peak Power
- Transient Power
- Cycle Difference Power
- Peak Power Differential
- Cycle-to-Cycle Power Gradient (Fluctuation)
- and many more





Why peak power reduction ?

- To maintain supply voltage levels
- To increase reliability
- To use smaller heat sinks
- To make packaging cheaper





Why Average Power/ Energy reduction ?

- To increase battery life time
- To enhance noise margin
- To reduce energy costs
- To reduce use of natural resources
- To increase system reliability





Why Transience / Fluctuation Minimization ?

- To reduce power supply noise
- To reduce cross-talk and electromagnetic noise
- To increase battery efficiency
- To increase reliability





Low-power design: Key Principles

>Using the lowest possible supply voltage.

- Using the smallest geometry, highest frequency devices, but operating them at lowest possible frequency.
- Using parallelism and pipelining to lower required frequency of operation.
- Power management by disconnecting the power source when the system is idle.





Voltage, Frequency and Power Trade-offs

- Reduce Supply Voltage (V_{dd}): delay increases; performance degradation
- Reduce Clock Frequency (f): only power saving no energy
- Reduce Switching Activity (N or E(sw)): no switching no power loss !!! Not in fully under designers control. Switching activity depends on the logic function. Temporal/and spatial correlations difficult to handle.
- Reduce Physical Capacitance: done by reducing device size reduces the current drive of the transistor making the circuit slow





How Much is Saved ?? Varying V_{dd} / f

Voltage (V _{dd})	Frequency (f)	Power (P _d)	Energy (E _d)
V _{dd}	f _{max}	P _d	E _d
V _{dd} / 2	f _{max} *	P _d / 4	E _d / 4
V _{dd} / 2	f _{max} / 2	P _d / 8	E _d / 4
V _{dd}	f _{max} / 2	P _d / 2	E _d

* Note :
$$f_{max}$$
 Vs f





Low Power Design : Static Reduction

- Reduce static power:
 - Selectively use ratioed circuits.
 - Selectively use low V_{th} devices.
 - Leakage reduction:

Stacked devices, body bias, low temperature.



